

FPGA Design Techniques

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Objectives

After completing this module, you will be able to:

- Increase design performance by duplicating flip-flops
- Increase design performance by adding pipeline stages
- · Increase board performance by using I/O flip-flops
- Build reliable synchronization circuits



Outline

- Duplicating Flip-flops
 - · Pipelining
 - · I/O Flip-flops
 - Synchronization Circuits
 - · Summary

Duplicating Flip-Flops

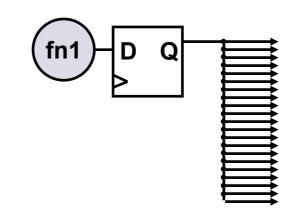
High-fanout nets can be slow and hard to route

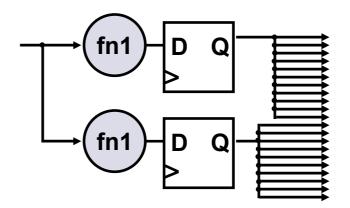
Duplicating flip-flops can fix both problems

- Reduced fanout shortens net delays
- Each flip-flop can fanout to a different physical region of the chip to help with routing

Design tradeoffs

- Gain routability and performance
- Design area increases







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Tips on Duplicating Flip-Flops

Name duplicated flip-flops _a, _b: NOT _1, _2

- Numbered flip-flops are mapped by default into the same slice
- You generally want duplicated flip-flops to be separated
 - Especially if the loads are spread across the chip

Explicitly create duplicate flip-flops in your HDL code

- Most synthesis tools have automatic fanout-control features
 - · However, they do not always pick the best division of loads
 - Also, duplicated flip-flops will be named _1, _2
- Many synthesis tools will optimize-out duplicated flip-flops
 - Tell your synthesis tool to keep redundant logic

Do not duplicate flip-flops that are sourced by asynchronous signals

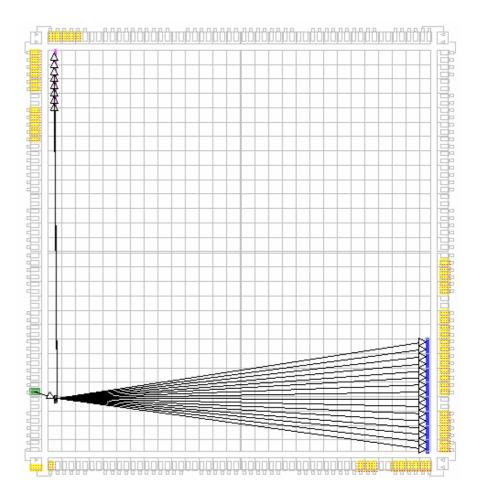
- Synchronize the signal first
- Feed synchronized signal to multiple flip-flops

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Duplicating Flip-Flops Example

- Source flip-flop drives two register banks that are constrained to different regions of the chip
- Source flip-flop and pad are not constrained
- PERIOD = 5 ns timing constraint
- Implemented with default options
- Longest path = 6.806 ns
 - Fails to meet timing constraint

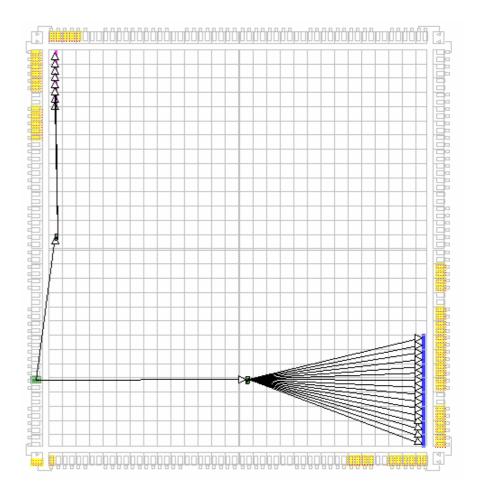


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Duplicating Flip-Flops Example

- Source flip-flop has been duplicated
- Each flip-flop drives a region of the chip
 - Each flip-flop can be placed closer to the register that it is driving
 - Shorter routing delays
- Longest path = 4.666 ns
 - Meets timing constraint





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Outline

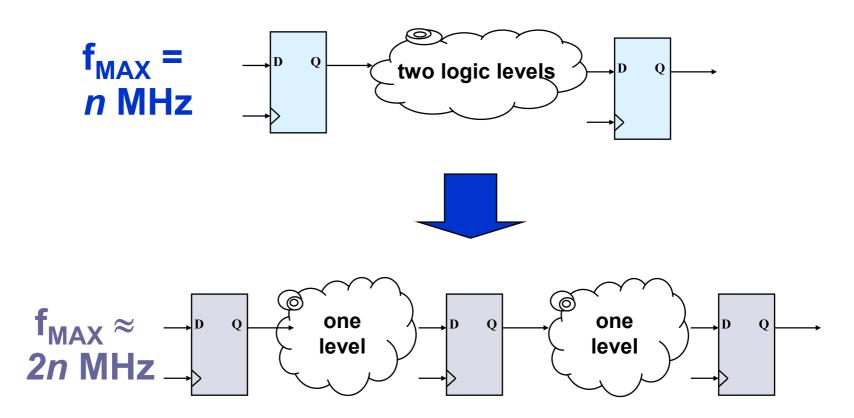
Duplicating Flip-Flops



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Pipelining Concept



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Pipelining Considerations

Are there enough flip-flops available?

- Refer to the Map Report
- You generally will not run out of FFS

Are there multiple logic levels between flip-flops?

- If there is only one logic level between flip-flops, pipelining will not improve performance
 - Exception: Long carry-logic chains can benefit from pipelining
- Refer to the Post-Map Static Timing Report or Post-Place & Route Static Timing Report

Can the system tolerate *latency*?

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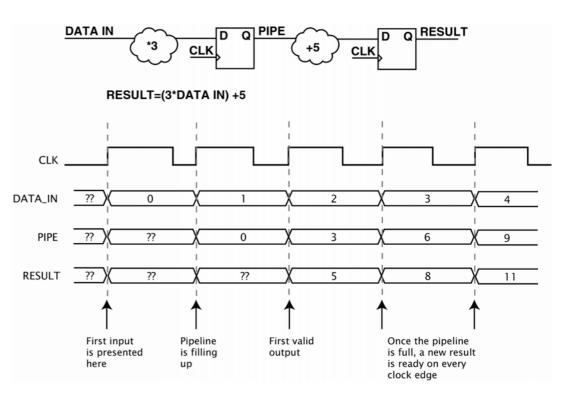
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Latency in Pipelines

Each pipeline stage adds one clock cycle of delay before the first output will be available

 Also called "filling the pipeline"

After the pipeline is filled, a new output is available every clock cycle



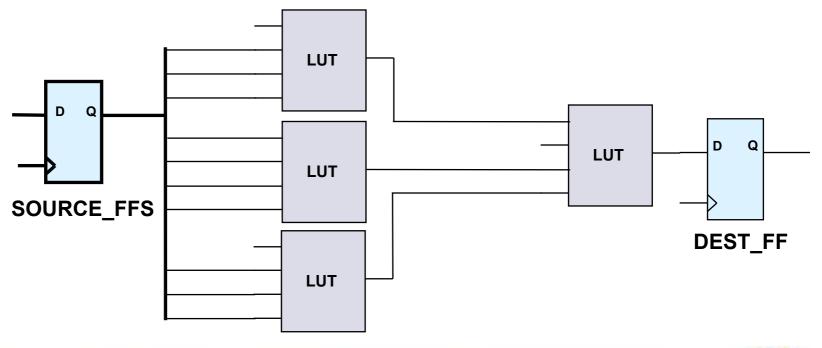


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Pipelining Example

Original circuit

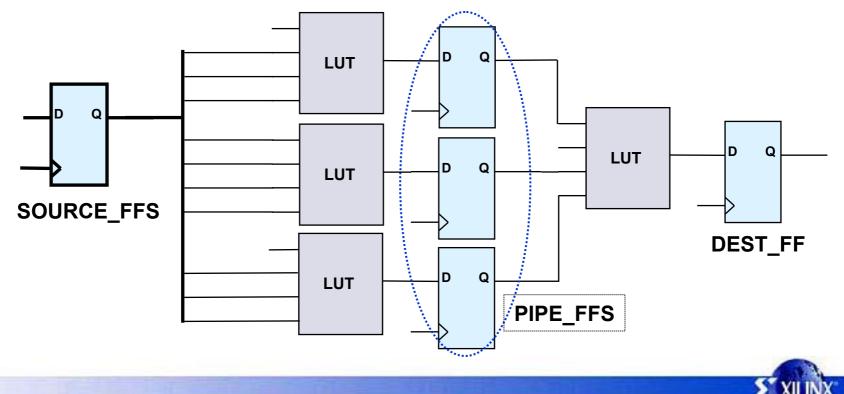
- Two logic levels between SOURCE_FFS and DEST_FF
- f_{MAX} = ∼207 MHz



Pipelining Example

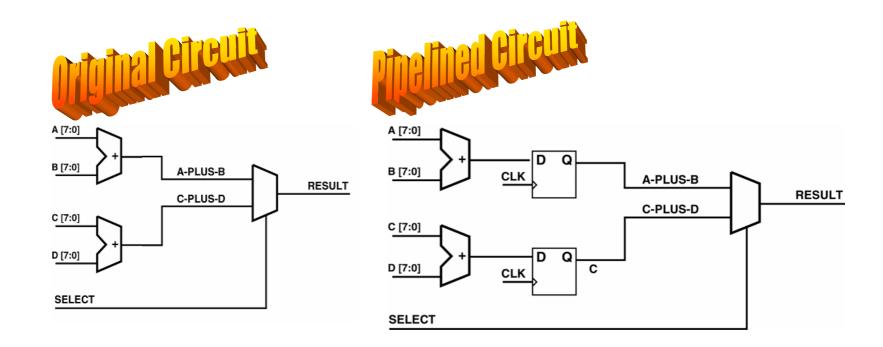
Pipelined circuit

- One logic level between each set of flip-flops
- f_{MAX} = ~347 MHz



Pipelining Questions

Given the original circuit, what is wrong with the pipelined circuit? How can the problem be corrected?



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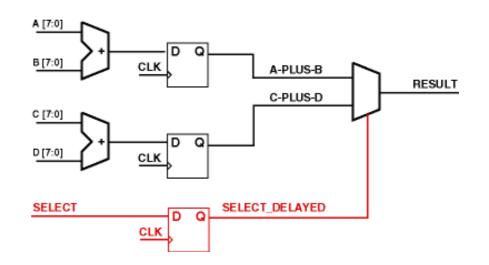
Pipelining Answers

What is wrong with the pipelined circuit?

- Latency mismatch
- Older data is mixed with newer data
- Circuit output is incorrect

How can the problem be corrected?

- Add a flip-flop on SELECT
- All data inputs now experience the same amount of latency





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- · Pipelining

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I/O Flip-Flop Overview

Each Virtex[™]-II I/O Block contains six flip-flops

- IFD on the input, OFD on the output, ENBFF on the 3-state enable
- Single or Double Data Rate support

Benefits of using I/O flip-flops

- Guaranteed setup, hold, and clock-to-out times when the clock signal comes from a BUFG
- Programmable delay on input flip-flop
- Programmable slew rate on output flip-flop



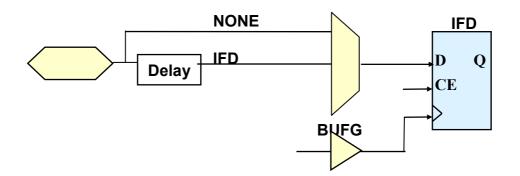
Programmable Input Delay

Delay is added to the D input of the IFD or ILD to guarantee 0 ns hold time

- Trades an increase in setup time for a decrease in hold time

Delay is controlled by the IOBDELAY attribute

- NONE is the default if CLK is driven by a DCM
- IFD is the default if CLK is not driven by a DCM



Programmable Input Delay Example

XC2V250-5 (v1.90 Virtex[™]-II speed files)

- CLK is a DCM output (default delay is NONE)
 - . IFD t_{SU} = 4.0 ns t_{H} = 0.0 ns
 - $\cdot \text{ NONE } t_{SU} = 1.3 \text{ ns} \qquad t_{H} = 0.0 \text{ ns}$
- CLK is not a DCM output (default delay is IFD)
 - $\cdot \quad \text{IFD} \qquad t_{\text{SU}} = 1.2 \text{ ns} \qquad \qquad t_{\text{H}} = 0.0 \text{ ns}$
 - NONE t_{SU} = -1.5 ns t_{H} = 1.7 ns

Programmable Output Slew Rate

Only the LVTTL and LVCMOS I/O standards support programmable slew rate

Slew rate can be SLOW or FAST

- By default, the slew rate is SLOW
- Slew rate is controlled by attributes
 - Specified in the Constraints Editor or the UCF file



Accessing I/O Flip-Flops

During synthesis

- Timing-driven synthesis can force flip-flops into IOBs
- Some tools support attributes or synthesis directives to mark flip-flops for placement in an IOB

Xilinx Constraint Editor

- Select the Misc tab
- Specify individual registers that should be placed in the IOB
 - · Need to know the instance name of each register

During the Map phase of implementation

- Map Properties, Pack I/O Registers/Latches into IOBs is on by default
- Timing-driven packing will also move registers into IOBs for critical paths

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Synchronization Circuits

What is a synchronization circuit?

- Captures an asynchronous input signal and outputs it on a clock edge

Why do I need synchronization circuits?

- Prevent setup and hold time violations
- End result is a more reliable design

When do I need synchronization circuits?

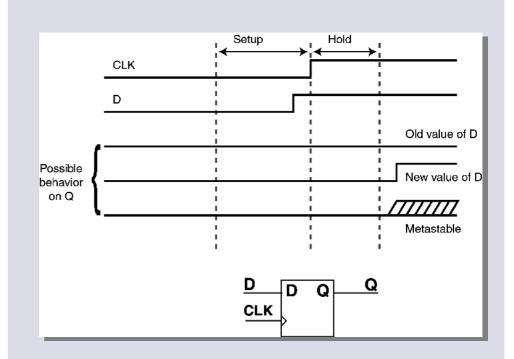
- Signals cross between *unrelated* clock domains
 - · Between related clock domains, a path-specific timing constraint is sufficient
- Chip inputs that are asynchronous

Setup and Hold Time Violations

Violations occur when the flip-flop input changes too close to a clock edge

Three possible results:

- Flip-flop clocks in old data value
- Flip-flop clocks in new data value
- Flip-flop output becomes *metastable*





Metastability

Flip-flop output enters a transitory state

- Neither a valid 0 nor a valid 1
 - May be interpreted as 0 by some loads and as 1 by others
- Remains in this state for an unpredictable length of time before settling to a valid 0 or 1
- Due to its statistical nature, the occurrence of metastable events can only be reduced, not eliminated
- Mean Time Between Failures (MTBF) is exponentially related to the length of time the flip-flop is given to recover
 - A few extra ns of recovery time can dramatically reduce the chances of a metastable event

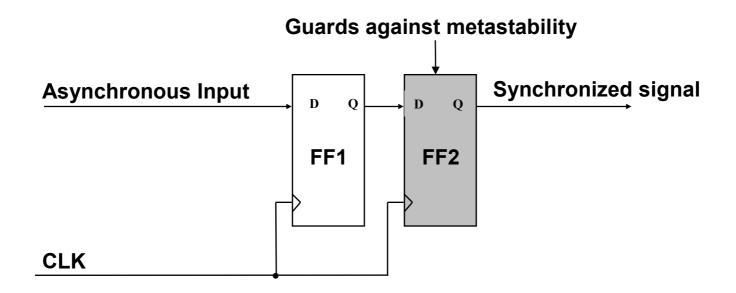
The circuits shown in this section allow a full clock cycle for metastable recovery

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Synchronization Circuit #1

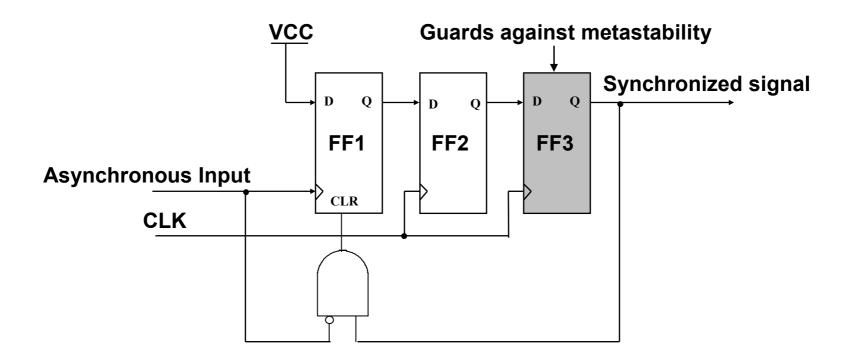
Use when input pulses will always be at least one clock period wide The "extra" flip-flops guard against metastability



Synchronization Circuit #2

Use when input pulses may be less than one clock period wide

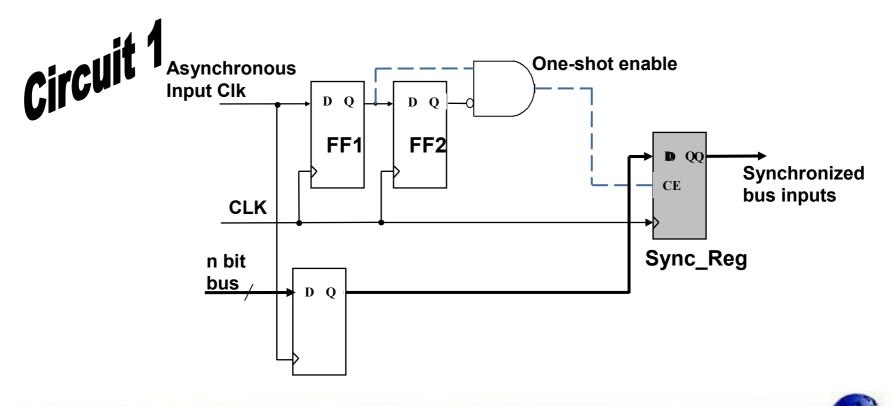
- FF1 captures short pulses



Capturing a Bus

Leading edge detector

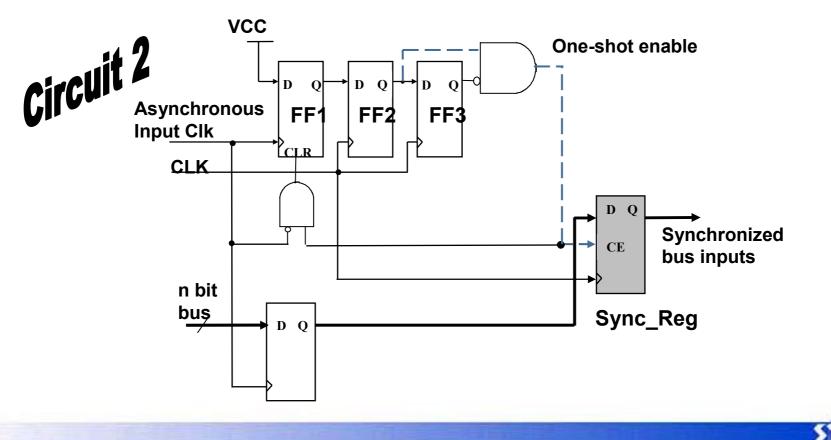
- Input pulses must be at least one CLK period wide



Capturing a Bus

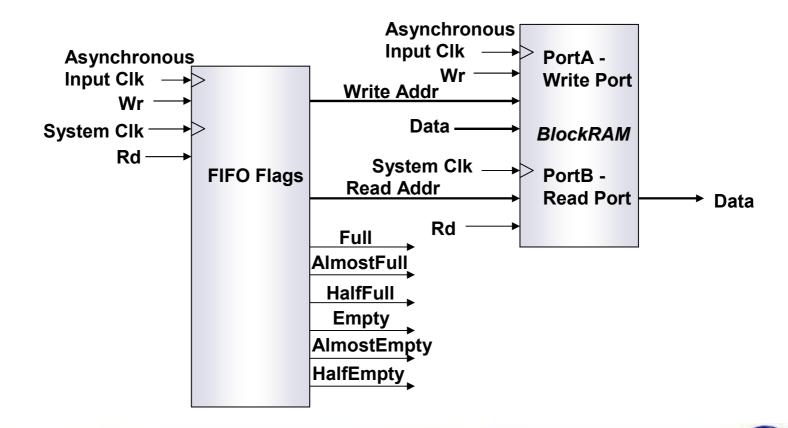
Leading edge detector

Input pulses may be less than one CLK period wide



Synchronization Circuit #3

Use a FIFO to cross domains



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Review Questions

High fanout is one reason to duplicate a flip-flop. What is another reason?

Give an example of a time when you do not need to resynchronize a signal that crosses between clock domains

What is the purpose of the "extra" flip-flop in the synchronization circuits shown in this module?

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Answers

High fanout is one reason to duplicate a flip-flop. What is another reason?

- Loads are divided between multiple locations on the chip

Give an example of a time when you do not need to resynchronize a signal that crosses between clock domains

- Well-defined phase relationship between the clocks
- Example: Clocks are the same frequency, 180° out of phase
- Use a path-specific constraint to ensure that data paths will meet timing

What is the purpose of the "extra" flip-flop in the synchronization circuits shown in this module?

- To allow the first flip-flop time to recover from metastability

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Summary

You can increase circuit performance by:

- Duplicating flip-flops
- Adding pipeline stages
- Using I/O flip-flops

Some tradeoffs

- Duplicating flip-flops increases circuit area
- Pipelining introduces latency
- I/O flip-flop programmable delay trades off setup time and hold time

Synchronization circuits increase reliability

Where Can I Learn More?

Data Book: http://support.xilinx.com \rightarrow Documentation \rightarrow Databook

- Switching Characteristics
- Detailed Functional Description > Input/Output Blocks (IOBs)
- Application Notes: http://support.xilinx.com \rightarrow Documentation \rightarrow App Notes
 - XAPP094: Metastability Recovery
 - XAPP225: Data to Clock Phase Alignment

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